

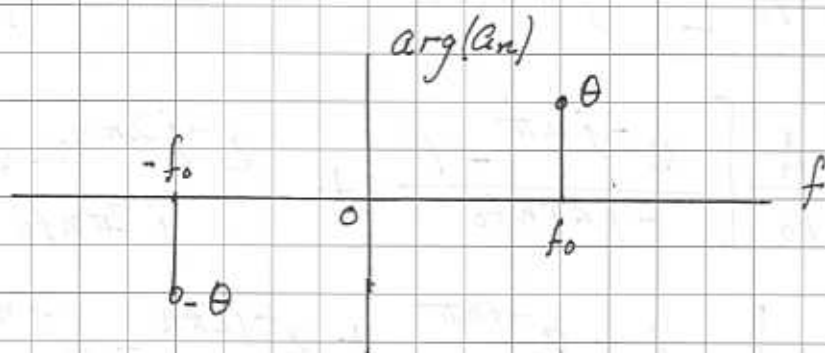
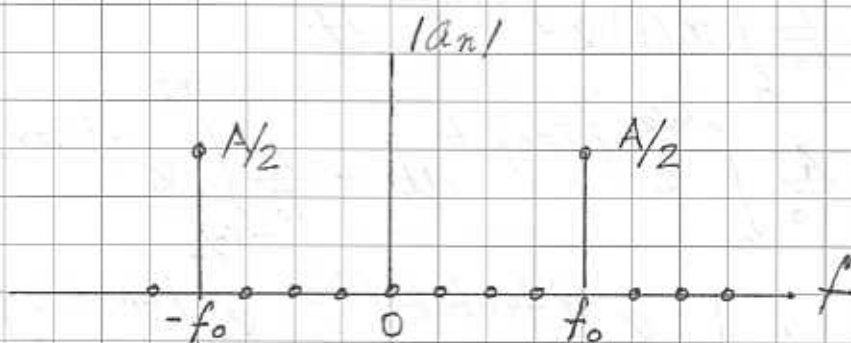
DIGITAL STORAGE OSCILLOSCOPE, FUNCTION GENERATOR, & MEASUREMENTS

PRELABSPECTRUM OF SINUSOID: assume $dc=0$ to start

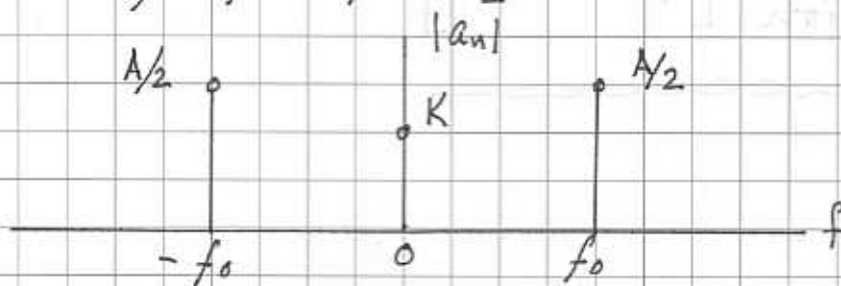
$$x(t) = A \cos(2\pi f_0 t + \theta)$$

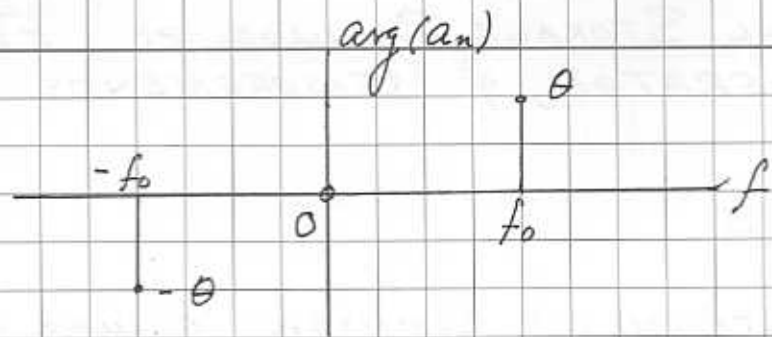
$$= \frac{A}{2} e^{j\theta} e^{j2\pi f_0 t} + \frac{A}{2} e^{-j\theta} e^{-j2\pi f_0 t}$$

$$\Rightarrow a_1 = \frac{A}{2} e^{j\theta}, \quad a_{-1} = \frac{A}{2} e^{-j\theta} = a_1^*$$

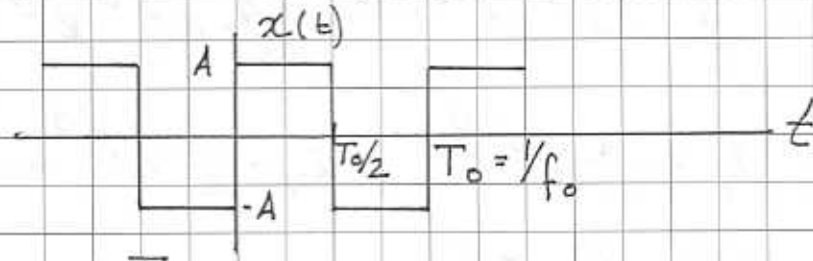
all other $a_n = 0$.Now for non-zero dc: $x(t) = K + A \cos(2\pi f_0 t + \theta)$

$$a_0 = K, \quad a_1 = a_{-1}^* = \frac{A}{2} e^{j\theta}$$





SPECTRUM OF SQUARE WAVE: Assume $dc=0$ first.



$$a_n = \frac{1}{T_0} \int_0^{T_0} x(t) e^{-j2\pi n f_0 t} dt$$

$$= \frac{A}{T_0} \int_0^{T_0/2} e^{-j2\pi n f_0 t} dt - \frac{A}{T_0} \int_{T_0/2}^{T_0} e^{-j2\pi n f_0 t} dt$$

Provided $n \neq 0$

$$= \frac{A}{T_0} \left[\frac{e^{-j2\pi n f_0 t}}{-j2\pi n f_0} \Big|_0^{T_0/2} + \frac{e^{-j2\pi n f_0 t}}{j2\pi n f_0} \Big|_{T_0/2}^{T_0} \right]$$

$$= \frac{A}{T_0} \left[\frac{e^{-jn\pi} - 1}{-j2\pi n f_0} + \frac{e^{-j2\pi n} - e^{-jn\pi}}{j2\pi n f_0} \right]$$

$$= A \left[\frac{1 - e^{-jn\pi} + e^{-j2\pi n} - e^{-jn\pi}}{j2\pi n} \right]$$

$$= \frac{A}{j\pi n} [1 - e^{-jn\pi}]$$

$$\text{For } n=0: a_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt = \underline{\underline{0}}$$

OR

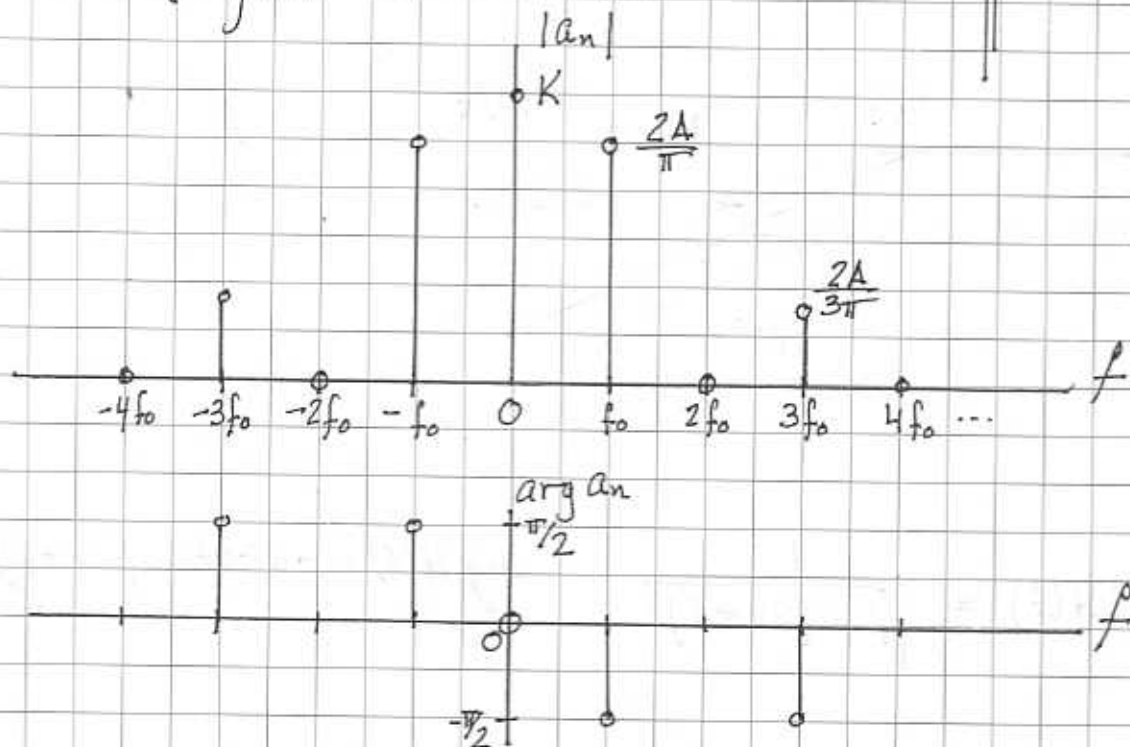
$$a_n = \begin{cases} 0 & (n=0) \\ \frac{A}{j\pi n} [1 - e^{-jn\pi}] & (n \neq 0) \end{cases} = \begin{cases} 0 & , n=0 \text{ or even} \\ \frac{2A}{j\pi n} & , n \text{ odd} \end{cases}$$

$n =$	0	1	2	3	4	-1	-2	-3	-4
$ a_n =$	0	$\frac{2A}{\pi}$	0	$\frac{2A}{3\pi}$	0	$\frac{2A}{\pi}$	0	$\frac{2A}{3\pi}$	0
$\arg(a_n) =$	-	$-\pi/2$	-	$-\pi/2$	-	$\pi/2$	-	$\pi/2$	-

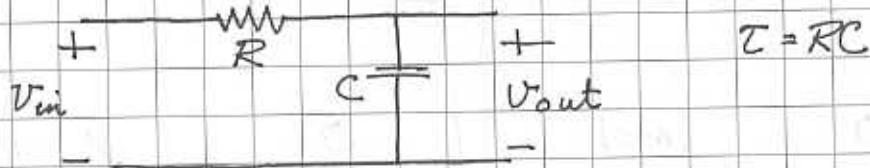
For a non-zero dc level K :

$$a_n = \begin{cases} K & , n=0 \\ 0 & , n \text{ even} \\ \frac{2A}{j\pi n} & , n \text{ odd} \end{cases}$$

F-coeff.
for a square
wave.



RC LOWPASS FILTER:



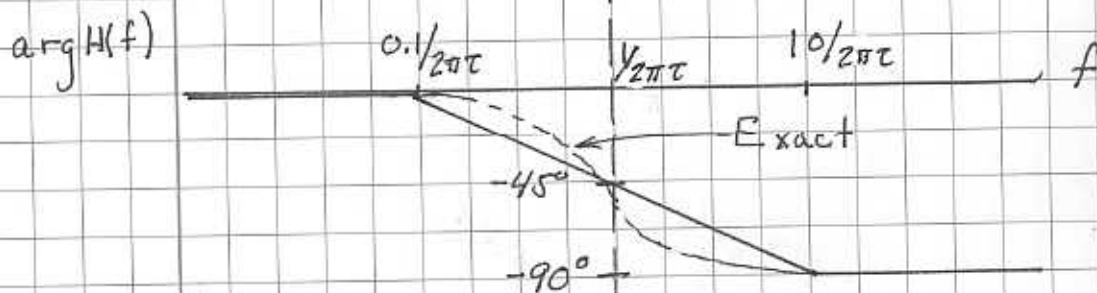
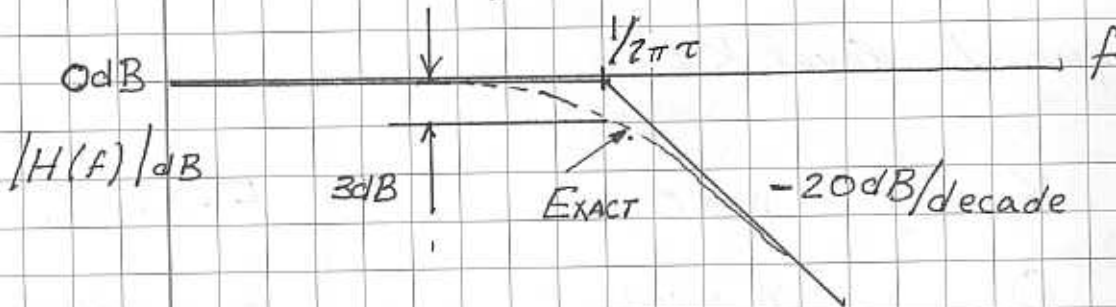
$$H(s) = \frac{V_{out}(s)}{V_{in}(s)} = \frac{1/Cs}{R + 1/Cs} = \frac{1}{1 + \tau s}, \quad \text{Re}(s) < -1/\tau$$

$$\underline{\underline{H(f) = \frac{1}{1 + j2\pi\tau f}}}$$

3dB break freq:

$$f_{3dB} = \frac{1}{2\pi\tau}$$

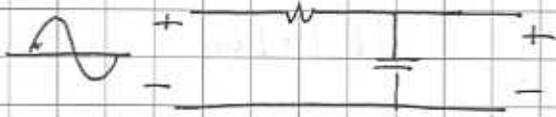
Bode Plots for $H(f)$:



$$|H(f)| = \frac{1}{\sqrt{1 + (2\pi\tau f)^2}}$$

$$\arg H(f) = -\arctan(2\pi\tau f)$$

FILTER RESPONSE TO SINUSOIDAL INPUT:



$$V_{in}(t) = K + A \cos(2\pi f_0 t + \theta)$$

$$a_0 = K, \quad a_1 = \frac{A}{2} e^{j\theta}, \quad a_{-1} = \frac{A}{2} e^{-j\theta}$$

~~$$H(f) = \frac{1}{1 + j2\pi\tau f}$$~~

$$H(f) = \frac{1}{1 + j2\pi\tau f}$$

$$H(0) = 1, \quad H(f_0) = \frac{1}{1 + j2\pi\tau f_0}, \quad H(-f_0) = \frac{1}{1 - j2\pi\tau f_0}$$

The output is sinusoidal with coefficients

$$b_n = a_n \cdot H(nf_0)$$

$$\Rightarrow b_0 = a_0 = K$$

$$b_1 = a_1 H(f_0) = \frac{A}{2} e^{j\theta} \frac{1}{1 + j2\pi\tau f_0}$$

$$= |b_1| e^{j\phi}$$

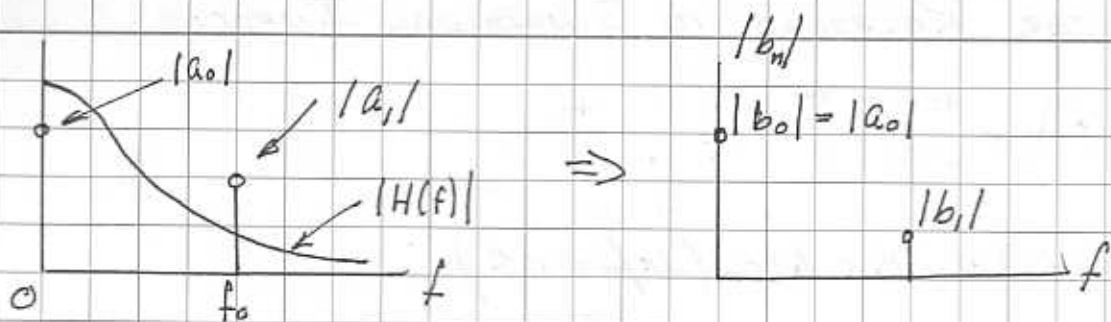
$$|b_1| = |a_1| \cdot |H(f_0)| = \frac{A}{2} \frac{1}{\sqrt{1 + (2\pi\tau f_0)^2}}$$

$$\phi = \arg a_1 + \arg H(f_0) = \theta - \arctan(2\pi\tau f_0)$$

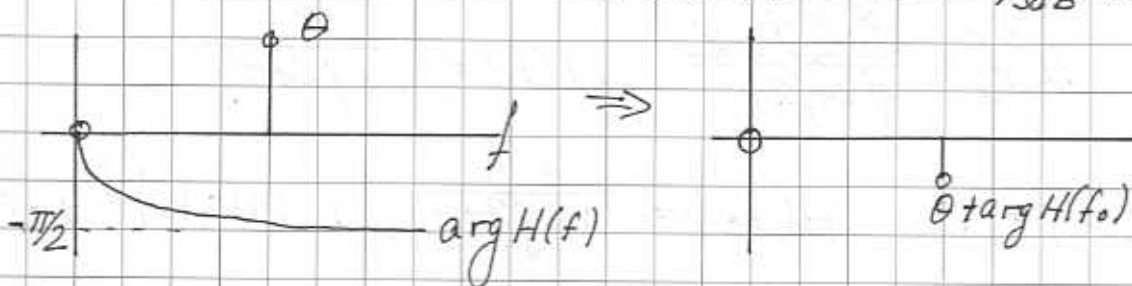
Like wise, $b_{-1} = b_1^*$

$$|b_{-1}| = |b_1|$$

$$\arg b_{-1} = -\arg b_1$$



Actual plot depends on relation between f_{3dB} and f_0



Output signal:

$$v_{out}(t) = b_0 + b_1 e^{j2\pi f_0 t} + b_{-1} e^{-j2\pi f_0 t}$$

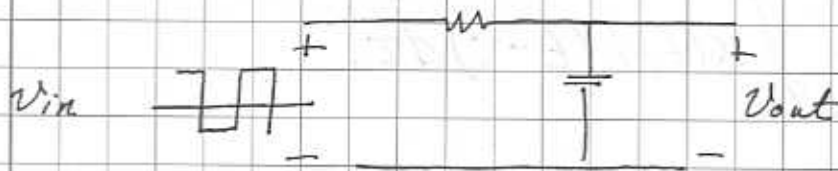
$$= b_0 + b_1 e^{j2\pi f_0 t} + (b_1 e^{j2\pi f_0 t})^*$$

$$= b_0 + 2\operatorname{Re}(b_1 e^{j2\pi f_0 t})$$

$$= b_0 + 2|b_1| \cos(2\pi f_0 t + \arg b_1)$$

$$= K + \frac{A}{\sqrt{1+(2\pi\tau f_0)^2}} \cos(2\pi f_0 t + \theta - \arctan 2\pi\tau f_0)$$

FILTER RESPONSE TO SQUARE WAVE INPUT



You could try to do the calculation in the frequency domain:

$$V_{in}(t) = \sum_{n=-\infty}^{\infty} A_n e^{j2\pi n f_0 t}, \quad A_n \text{ calculated on p. 19.}$$

$$\text{Then } V_{out}(t) = \sum_{n=-\infty}^{\infty} b_n e^{j2\pi n f_0 t}$$

$$b_n = A_n \cdot H(nf_0) = \begin{cases} K & , n=0 \\ 0 & , n \text{ even} \\ \frac{2A}{j\pi n} \cdot \frac{1}{1+j2\pi\tau n f_0} & , n \text{ odd} \end{cases}$$

Plot: $|a_n|$, arg a_n plot on p. 19, scaled by $|H(nf_0)|$ & phase shifted by arg $H(nf_0)$.

Then you must calculate $V_{out}(t)$ using

$$V_{out}(t) = \sum_{n=-N}^N b_n e^{j2\pi n f_0 t}$$

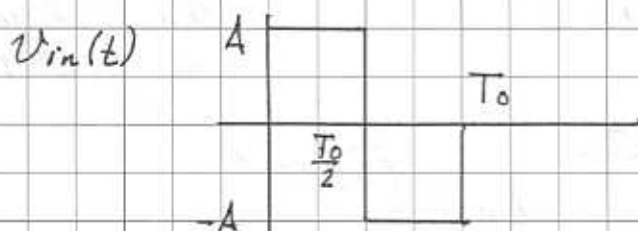
for large enough N . See Prelab 1. med for this method.

The output is easily calculated in the time domain by convolution: assume first $K=0$.

$$y(t) = \int_{-\infty}^{\infty} x(t) h(t-\tau) d\tau.$$

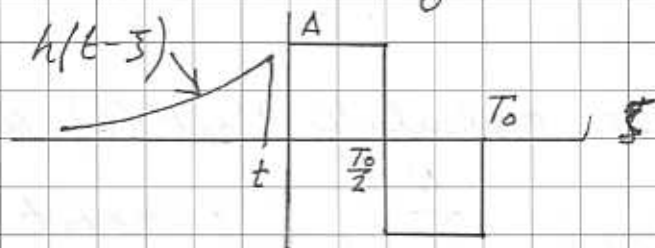
The input in this case is periodic, and so the output also is periodic with same period.

Calculate convolution of one period:



$$h(t) = \frac{1}{\tau} e^{-t/\tau} u(t)$$

$$\text{One period of } v_{out}(t) = \int_0^t v_{in}(\xi) h(t-\xi) d\xi$$



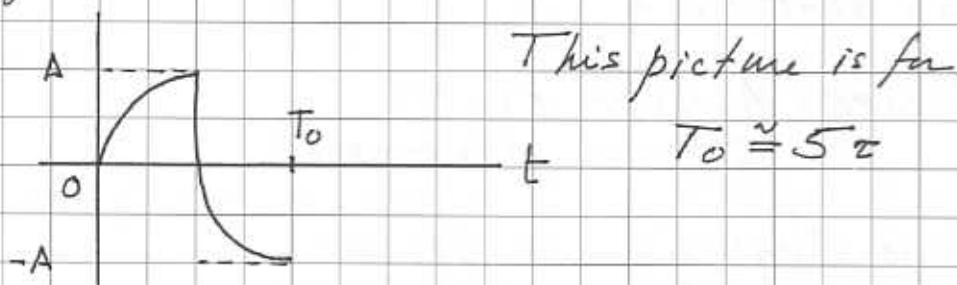
$$t < 0 \quad v_{out}(t) = 0$$

$$0 < t < T_0/2 \quad v_{out}(t) = A \int_0^t \frac{1}{\tau} e^{-(t-\xi)/\tau} d\xi = A(1 - e^{-t/\tau})$$

$$T_0/2 < t < T_0 \quad v_{out}(t) = A \int_0^{T_0/2} \frac{1}{\tau} e^{-(t-\xi)/\tau} d\xi - A \int_{T_0/2}^t \frac{1}{\tau} e^{-(t-\xi)/\tau} d\xi$$

$$= A(2e^{T_0/2\tau} - 1)e^{-t/\tau} - A$$

One period of $V_{out}(t)$:



For $K \neq 0$, we simply have a level shift by K in $V_{out}(t)$.

The convolution solution can also be done on the computer. See Prelab1.mcd.

RC filter with $\tau = 10 \mu\text{s}$ ($f_{3dB} = 15.92 \text{ kHz}$)

$$\tau = RC = 10^{-5}$$

Must use standard R & C values; also must have R large enough so as not to load down the function generator.

Some possibilities:

<u>R</u>	<u>C</u>
1k Ω	10nF
10k Ω	1nF

Will need to adjust these in lab according to available R and C .